

Lab number: 1

Lab title: TEM Wave in lossless media

Date lab was performed: 07.04.2020

Names of lab group members: Krzysztof Rudnicki

Theoretical introduction:

We are gonna perform simulation using rectangular dielectric slab with perfect electric conductor at the top and the bottom and perfect magnetic conductors at the lateral walls.

We know from boundary conditions that TEM is how electric polarization and corresponding magnetic component are gonna propagate.

Dielectric medium which fills the line is characterized by  $\epsilon_r$ ,  $\mu$  and  $tg\delta$ . Input port excites a sinusoidal TEM wave. Frequency ( $f$ ) is in GHz.

$a = 7.5$

1 a)

Values we start with:

$f = 7.5 [GHz]$ ,  $\epsilon_r = 1 [F / m]$ ,  $\mu_r = 1 [H / m]$ ,  $tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$  (lossless model)

$f$  – frequency,  $\epsilon_r$  – permittivity,  $\mu_r$  – permeability,  $tg\delta$  – loss tangent,  $\sigma$  – conductivity

3.7:

$Z_c \text{ of input} = Z_c \text{ of output} = 376.7303 [\Omega]$

$Z_c$  – impedance

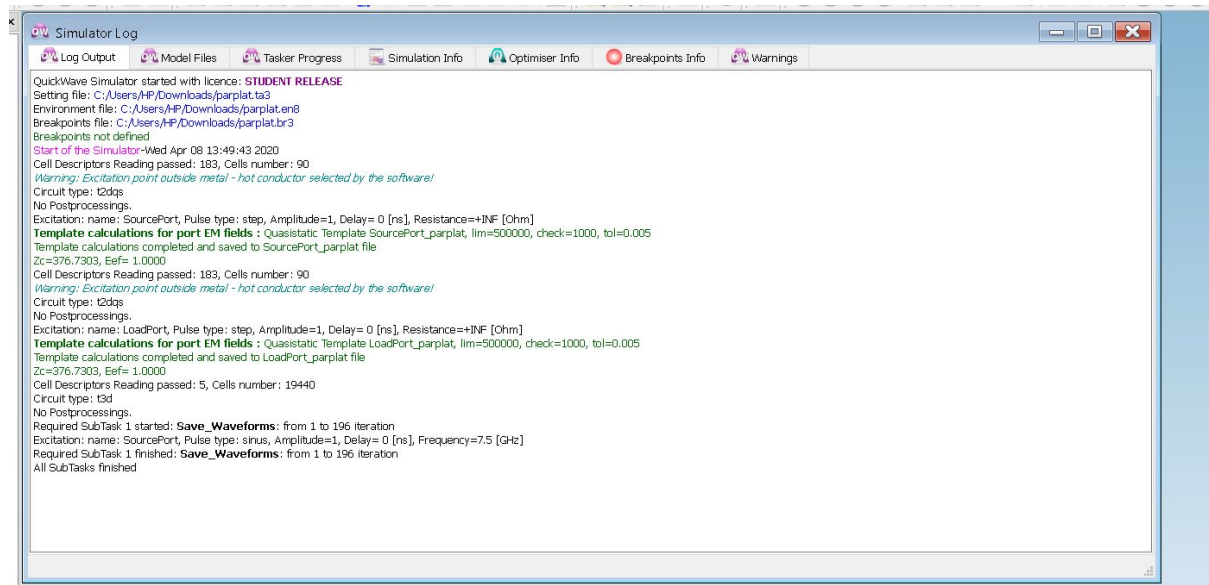


Figure 1: Impedance value for

$f = 7.5 [GHz]$ ,  $\epsilon_r = 1 [F / m]$ ,  $\mu_r = 1 [H / m]$ ,  $tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$

3.8:

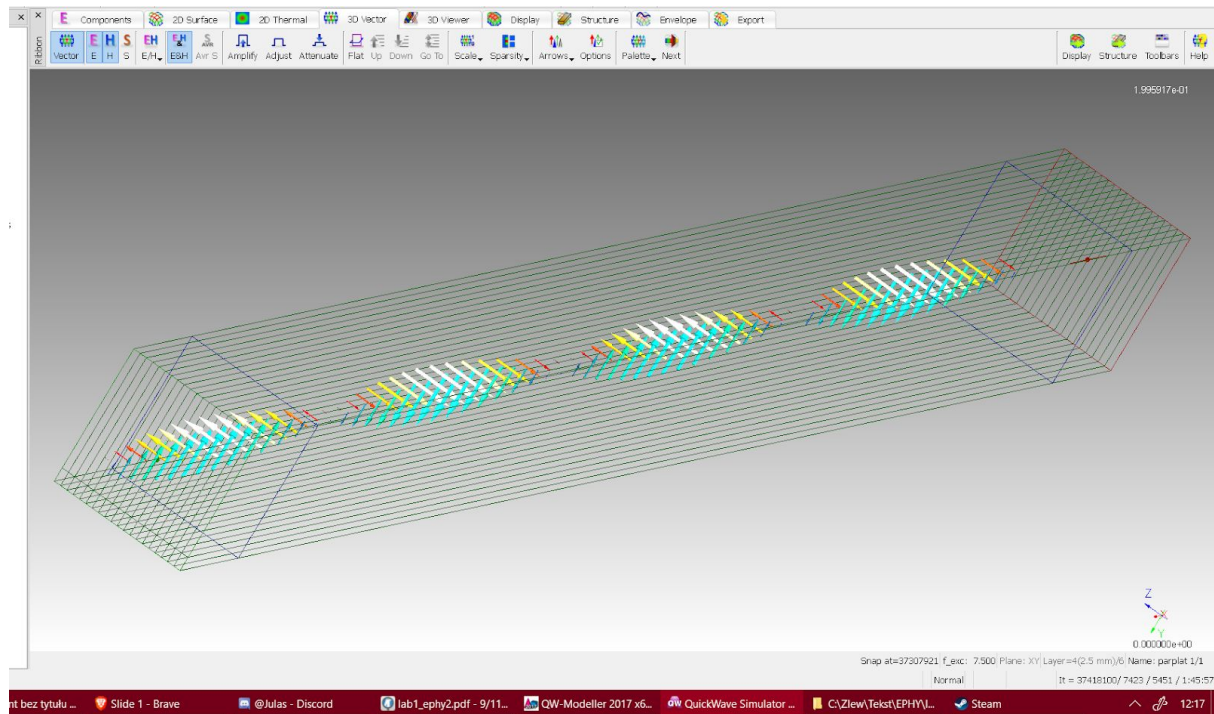


Figure 2: 3D Vector Display for

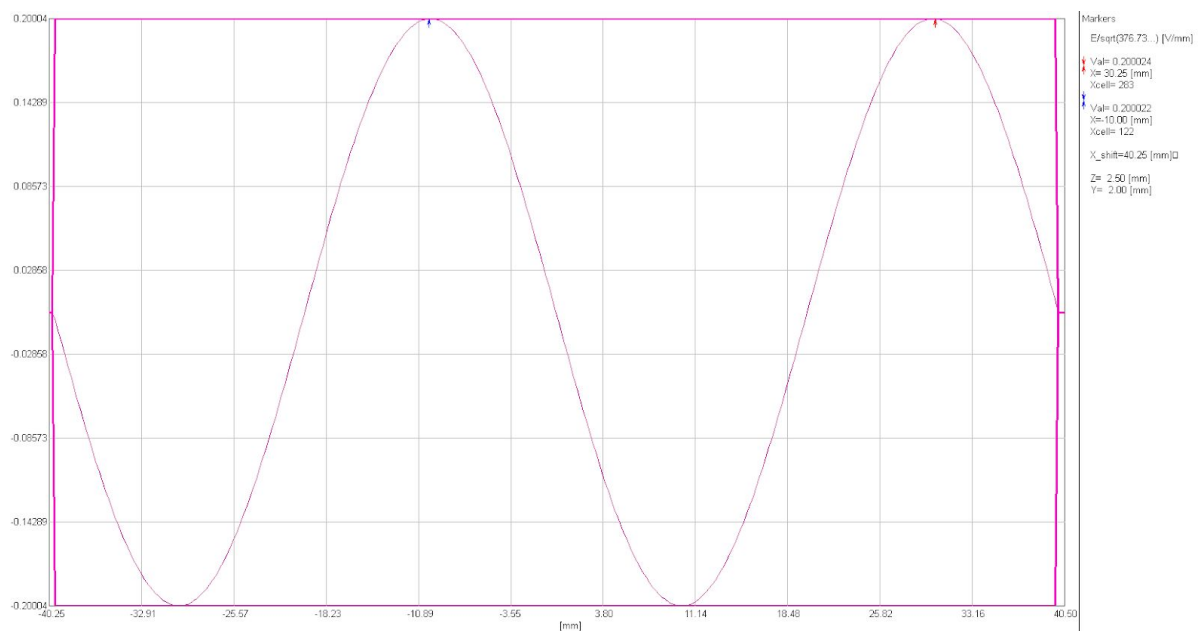
$$f = 7.5 \text{ [GHz]}, \epsilon_r = 1 \text{ [F/m]}, \mu_r = 1 \text{ [H/m]}, \text{tg}\delta = 0 \Rightarrow \sigma = 0 \text{ [S/m]}$$

Electric component: z (We can see that only z axis is visible in electric component in the program)

Magnetic component: y (We can see that only y axis is visible in magnetic component in the program)

Direction of propagation: x (It is the only axis left, after we established that we can see z and y component)

3.9



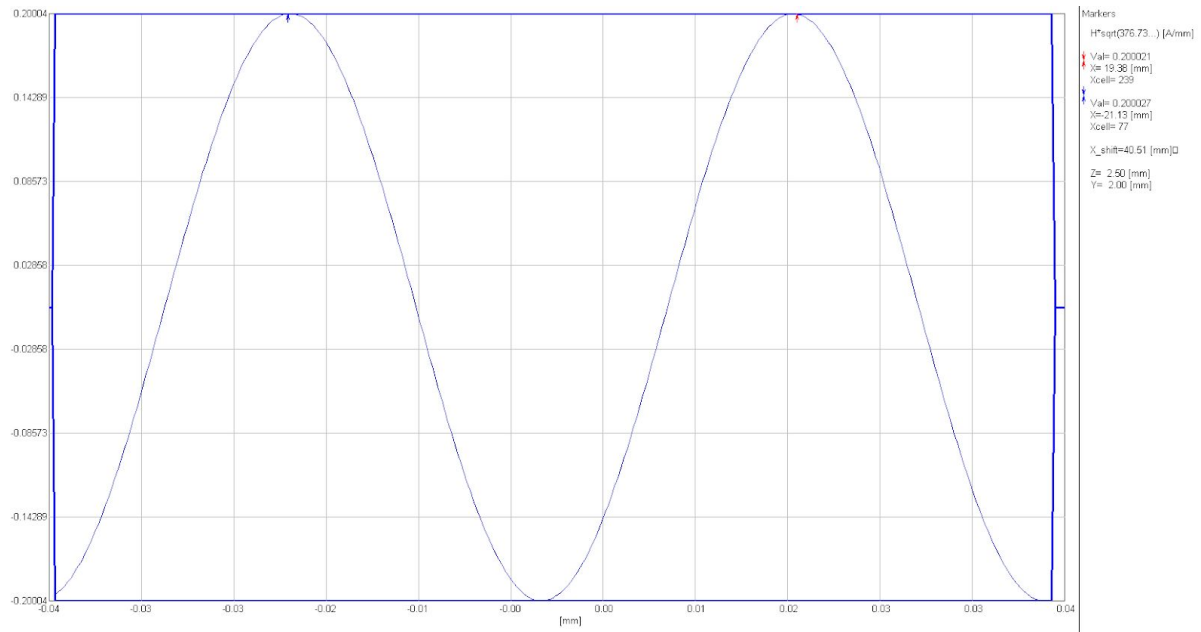


Figure 3: Envelope windows Ez(upper) Hy(lower) for  $f = 7.5 \text{ [GHz]}$ ,  $\epsilon_r = 1 \text{ [F / m]}$ ,  $\mu_r = 1 \text{ [H / m]}$ ,  $\text{tg}\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$

Wavelength -  $\lambda = X\_shift = 40.25 \text{ [mm]}$

Formula for phase coefficient  $\beta$  using measured  $\lambda$  :

$$\lambda = \frac{2\pi}{\beta} \Rightarrow \beta = \frac{2\pi}{\lambda} \Rightarrow \beta \approx 156.02 \text{ [1 / m]}$$

$$\epsilon = \epsilon_0 * \epsilon_r =$$

$$\text{Analytical formula for } \beta = \omega * \sqrt{\mu * \epsilon} = 2\pi * f * \sqrt{\mu * \epsilon} \approx 154,92 \text{ [1 / m]}$$

$\beta_{\text{markers}}$  - Phase coefficient calculated from lambda from markers

$\beta_{\text{analytical}}$  - Phase coefficient calculated from analytical formula

$$\text{Relative error} = 100 \% * \frac{\beta_{\text{markers}} - \beta_{\text{analytical}}}{\beta_{\text{analytical}}} \approx 0.7 \%$$

From markers:

$$E_n = 0.2 \text{ [V / mm]}$$

$$H_n = 0.2 \text{ [A / mm]}$$

$$E = E_n * \sqrt{Z_0} \approx 2.2 \text{ [V / mm]}$$

$$H = \frac{H_n}{\sqrt{Z_0}} = 0.018 \text{ [A / mm]}$$

$$Z_w = \frac{E_n}{H_n} * Z_0 = 376.7 \text{ [\Omega]}$$

From analytical formulas:

$$Z = \sqrt{\frac{\mu}{\epsilon}} = 377.0 \text{ [\Omega]}$$

$$\text{Relative error} = 100\% * \frac{Z - Z_w}{Z} \approx 0.08 \%$$

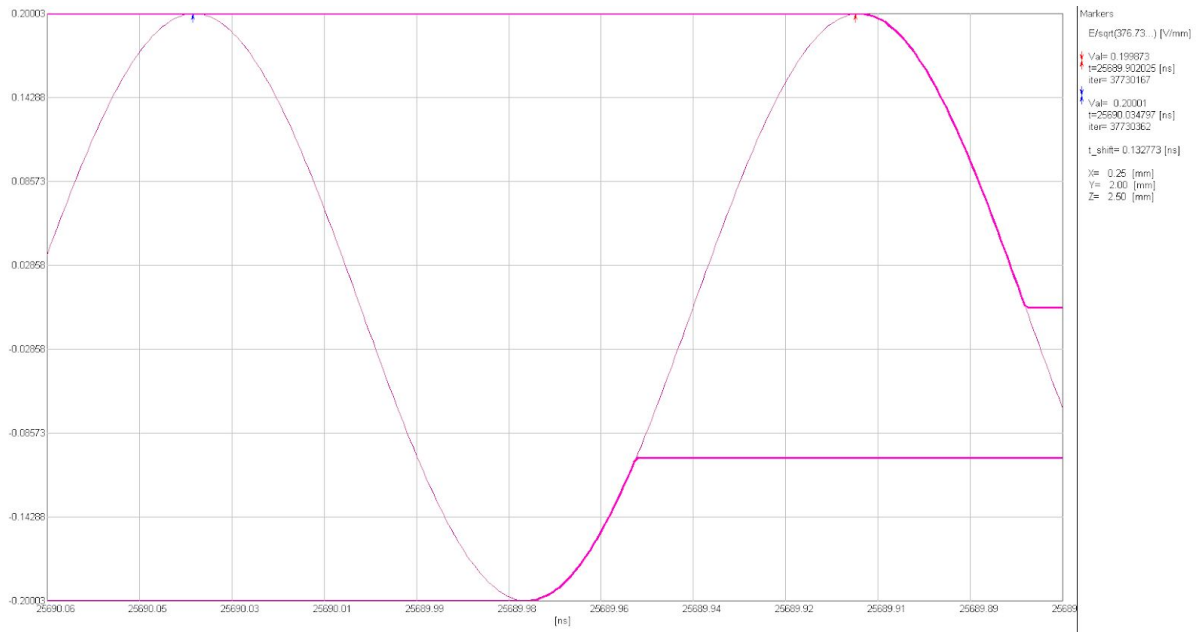


Figure 4: Time domain View Envelope window for  
 $f = 7.5 \text{ [GHz]}, \epsilon_r = 1 \text{ [F / m]}, \mu_r = 1 \text{ [H / m]}, tg\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$

$$T = t_{shift} \approx 0.138 \text{ [ns]}$$

$$T_{real} = \frac{1}{f_{real}} \approx 0.133 \text{ [ns]}$$

$$Relative\ error = \frac{T - T_{real}}{T_{real}} * 100\% = 3.3\%$$

$$f = \frac{1}{T} = 7.25 \text{ [GHz]}$$

$$f_{real} = 7.5 \text{ [GHz]}$$

$$Relative\ error = \frac{f_{real} - f}{f_{real}} * 100\% = 3.3\%$$

$$\beta = 2\pi * f * \sqrt{\mu * \epsilon} \approx 149.76 \text{ [1 / m]}$$

$$\beta_{analytical} = 154.92 \text{ [1 / m]}$$

$$Relative\ error = \frac{\beta_{analytical} - \beta}{\beta_{analytical}} * 100\% = 3.3\%$$

$\beta$  compared with  $\beta$  from 3.7 :

$$Relative\ error\ with\ \beta\ from\ 3.7 : \frac{\beta_{3.7} - \beta_{3.10}}{\beta_{3.7}} * 100\% \approx 3.31\%$$

$$a = 7.5$$

1 b)

Values we start with:

$$f = 15 \text{ [GHz]}, \epsilon_r = 1 \text{ [F / m]}, \mu_r = 1 \text{ [H / m]}, tg\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]} \text{ (lossless model)}$$

$f$  - frequency,  $\epsilon_r$  - permittivity,  $\mu_r$  - permeability,  $tg\delta$  - loss tangent,  $\sigma$  - conductivity

3.7:

$$Z_c\ of\ input = Z_c\ of\ output = 376.7303 \text{ [\Omega]}$$

$Z_c$  - impedance

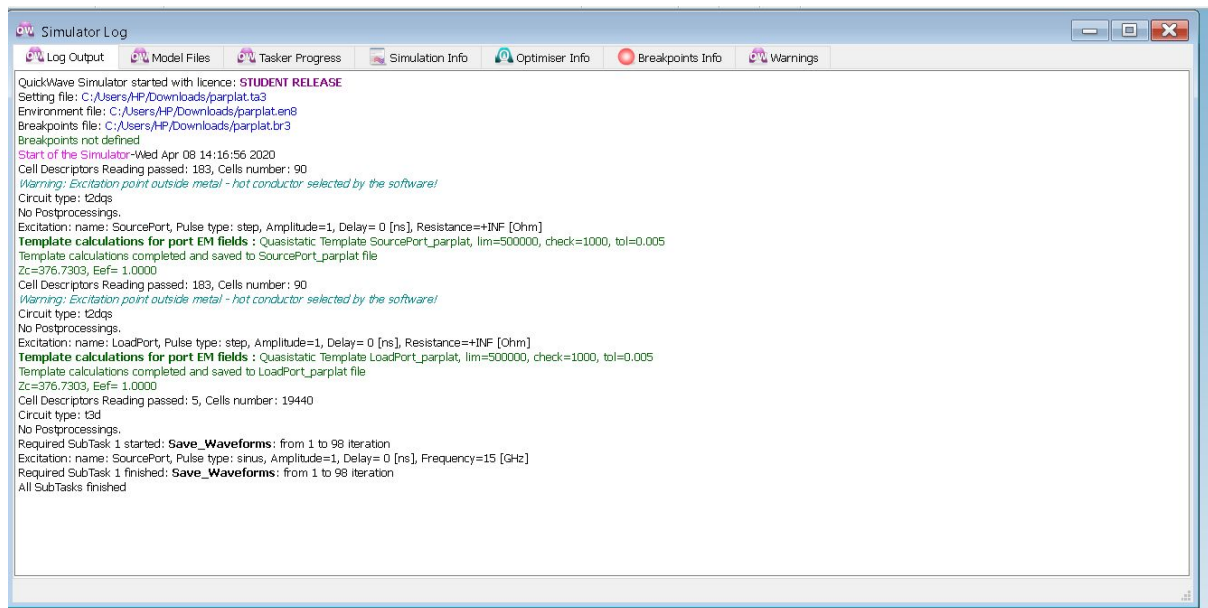


Figure 5: Impedance value for

$$f = 15 \text{ [GHz]}, \varepsilon_r = 1 \text{ [F / m]}, \mu_r = 1 \text{ [H / m]}, tg\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$$

3.8:

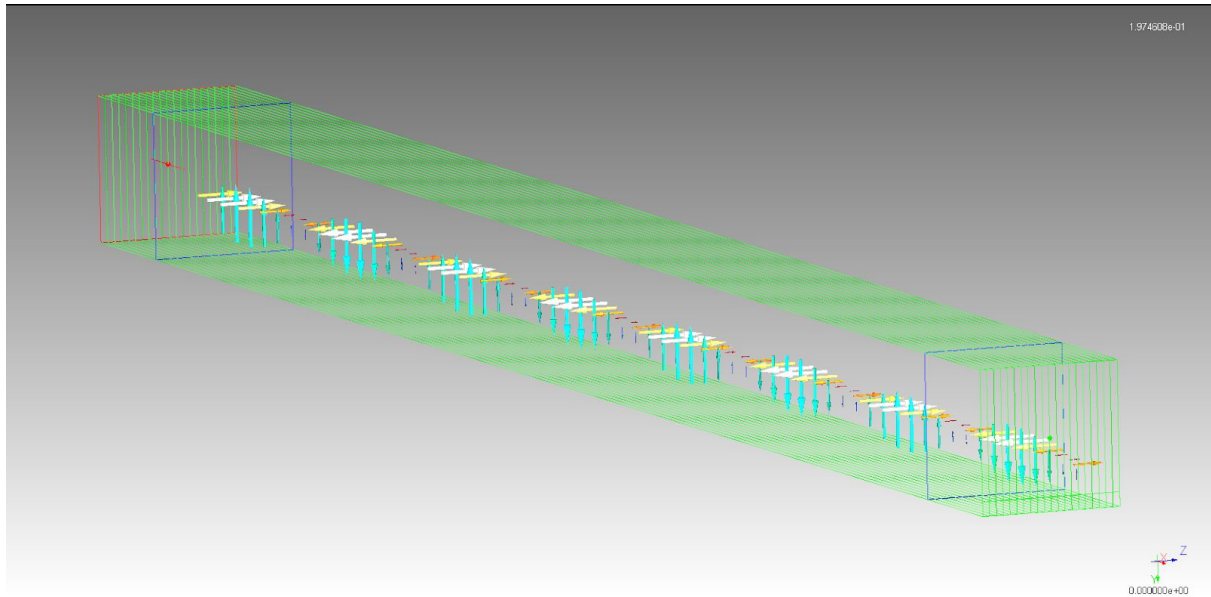


Figure 6: Fields window (3d vector display) for:

$$f = 15 \text{ [GHz]}, \epsilon_r = 1 \text{ [F / m]}, \mu_r = 1 \text{ [H / m]}, \text{tg}\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$$

Electric component: z (We can see that only z axis is visible in electric component in the program)

Magnetic component: y (We can see that only y axis is visible in magnetic component in the program )

Direction of propagation: x (It is the only axis left, after we established that we can see z and y component)

3.9

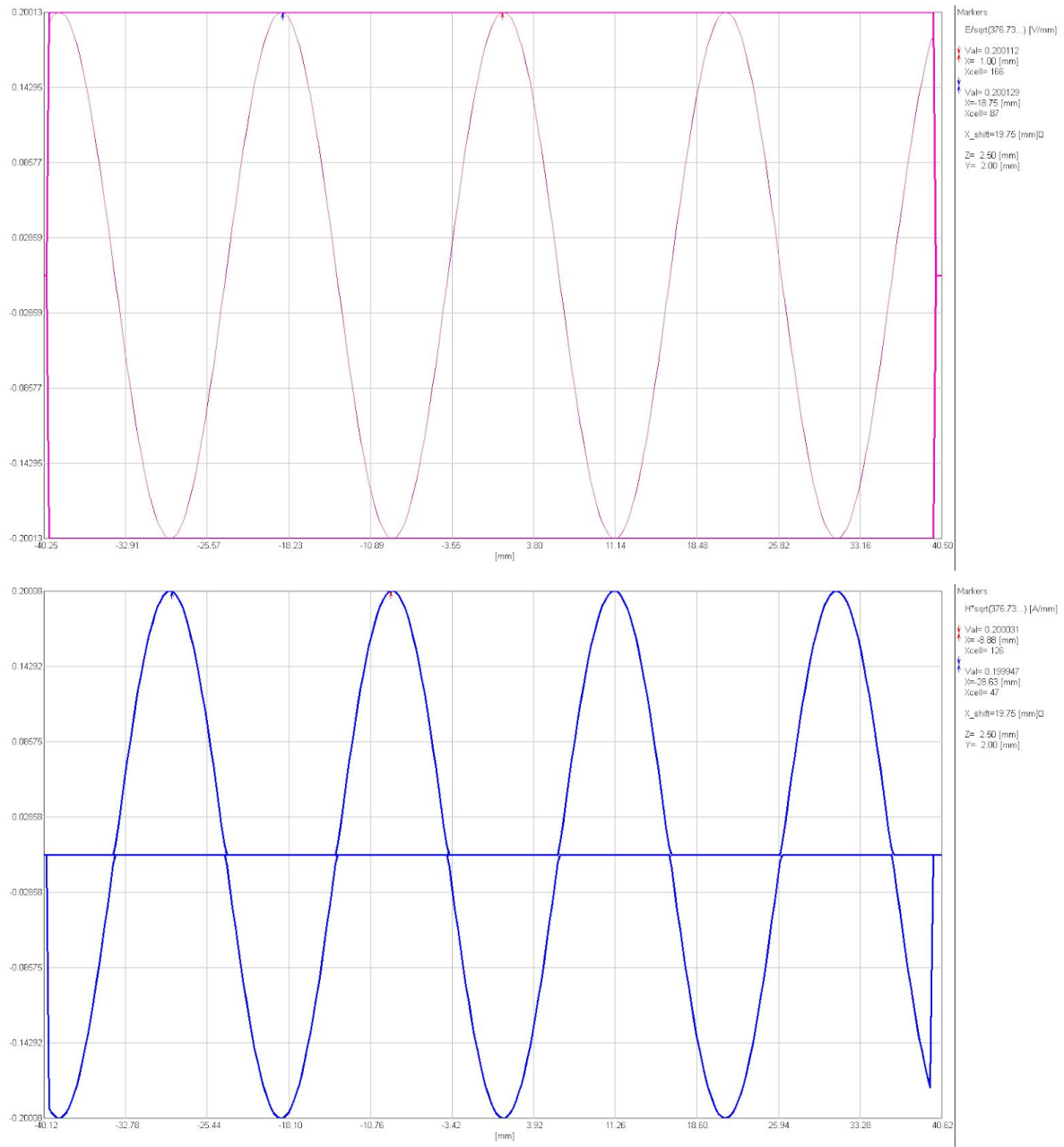


Figure 7: Envelope windows  $E_z$ (upper) and  $H_y$ (lower) for  $f = [GHz]$ ,  $\epsilon_r = 1 [F / m]$ ,  $\mu_r = 1 [H / m]$ ,  $tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$

Wavelength -  $\lambda = X\_shift = 19.75 [mm]$

Formula for phase coefficient  $\beta$  using measured  $\lambda$  :

$$\lambda = \frac{2\pi}{\beta} \Rightarrow \beta = \frac{2\pi}{\lambda} \Rightarrow \beta \approx 318.13 [1 / m]$$

$$\epsilon = \epsilon_0 * \epsilon_r$$

$$\text{Analytical formula for } \beta = \omega * \sqrt{\mu * \epsilon} = 2\pi * f * \sqrt{\mu * \epsilon} \approx 309.84 [1 / m]$$

$\beta_{markers}$  - Phase coefficient calculated from lambda from markers

$\beta_{analytical}$  - Phase coefficient calculated from analytical formula

$$\text{Relative error} = 100 \% * \frac{\beta_{markers} - \beta_{analytical}}{\beta_{analytical}} \approx 2.7 \%$$

From markers:



$$E_n = 0.2 [V / mm]$$

$$H_n = 0.2 [A / mm]$$

$$E = E_n * \sqrt{Z_0} \approx 2.2 [V / mm]$$

$$H = \frac{H_n}{\sqrt{Z_0}} = 0.018 [A / mm]$$

$$Z_w = \frac{E_n}{H_n} * Z_0 = 376.7 [\Omega]$$

From analytical formulas:

$$Z = \sqrt{\frac{\mu}{\epsilon}} = 377.0 [\Omega]$$

$$\text{Relative error} = 100\% * \frac{Z - Z_w}{Z} \approx 0.08 \%$$

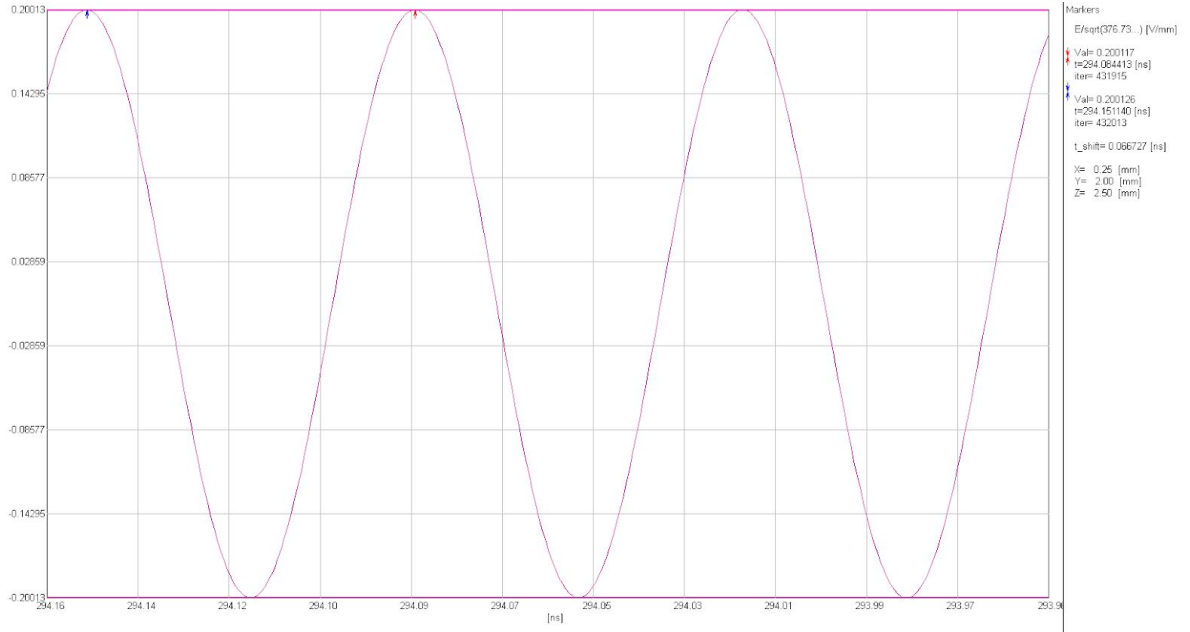


Figure 8: Time domain Ez component for:

$$f = 15 [GHz], \epsilon_r = 1 [F / m], \mu_r = 1 [H / m], tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$$

$$T = t_{shift} \approx 0.067 [ns]$$

$$T_{real} = \frac{1}{f_{real}} \approx 0.066 [ns]$$

$$\text{Relative error} = \frac{T - T_{real}}{T_{real}} * 100\% = 1.5\%$$

$$f = \frac{1}{T} = 14.93 [GHz]$$

$$f_{real} = 15 [GHz]$$

$$\text{Relative error} = \frac{f_{real} - f}{f_{real}} * 100\% = 0.46\%$$

$$\beta = 2\pi * f * \sqrt{\mu * \epsilon} \approx 308.4 [1 / m]$$

$$\beta_{analytical} = 309.84 [1 / m]$$

$$\text{Relative error} = \frac{\beta_{analytical} - \beta}{\beta_{analytical}} * 100\% = 0.46\%$$

$\beta$  compared with  $\beta$  from 3.7 :

$$\text{Relative error with } \beta \text{ from 3.7 : } \frac{\beta_{3.7} - \beta_{3.10}}{\beta_{3.7}} * 100\% \approx 3.05\%$$

2 b)

Values we start with:

$f = 7.5 \text{ [GHz]}, \epsilon_r = 1 \text{ [F / m]}, \mu_r = 7.5 \text{ [H / m]}, tg\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$  (lossless model)  
 $f$  – frequency,  $\epsilon_r$  – permittivity,  $\mu_r$  – permeability,  $tg\delta$  – loss tangent,  $\sigma$  – conductivity

3.7:

$Z_c \text{ of input} = Z_c \text{ of output} = Z_c = 1031.7184 \text{ [\Omega]}$

$Z_c$  – impedance

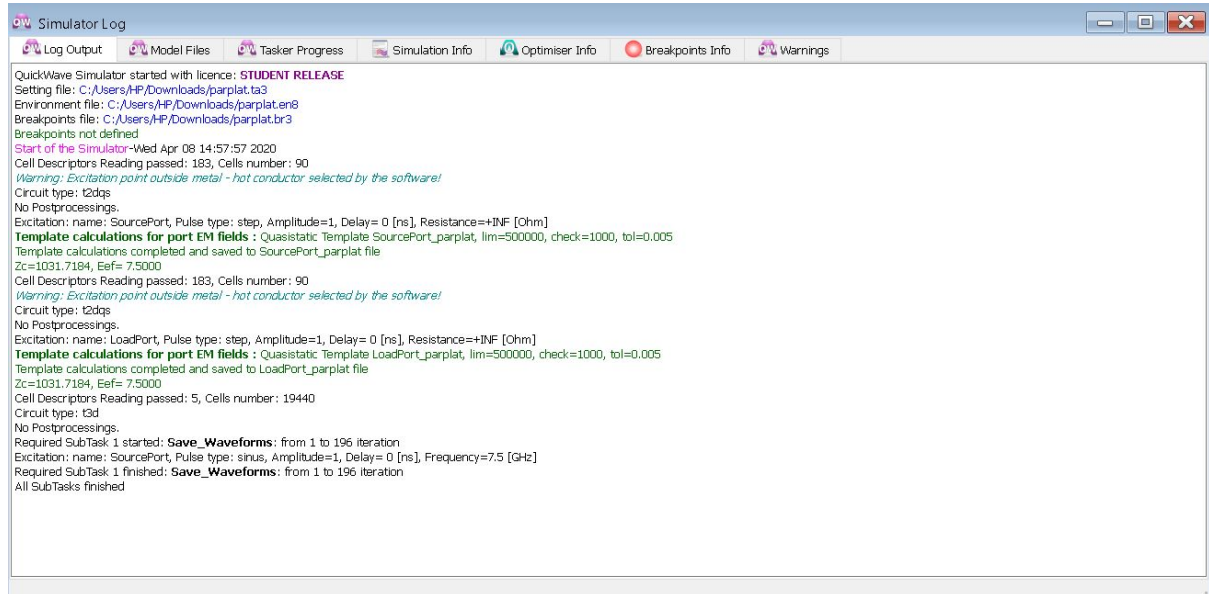


Figure 9: Simulation log window for

$f = 7.5 \text{ [GHz]}, \epsilon_r = 1 \text{ [F / m]}, \mu_r = 7.5 \text{ [H / m]}, tg\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$  (lossless model)

3.8:

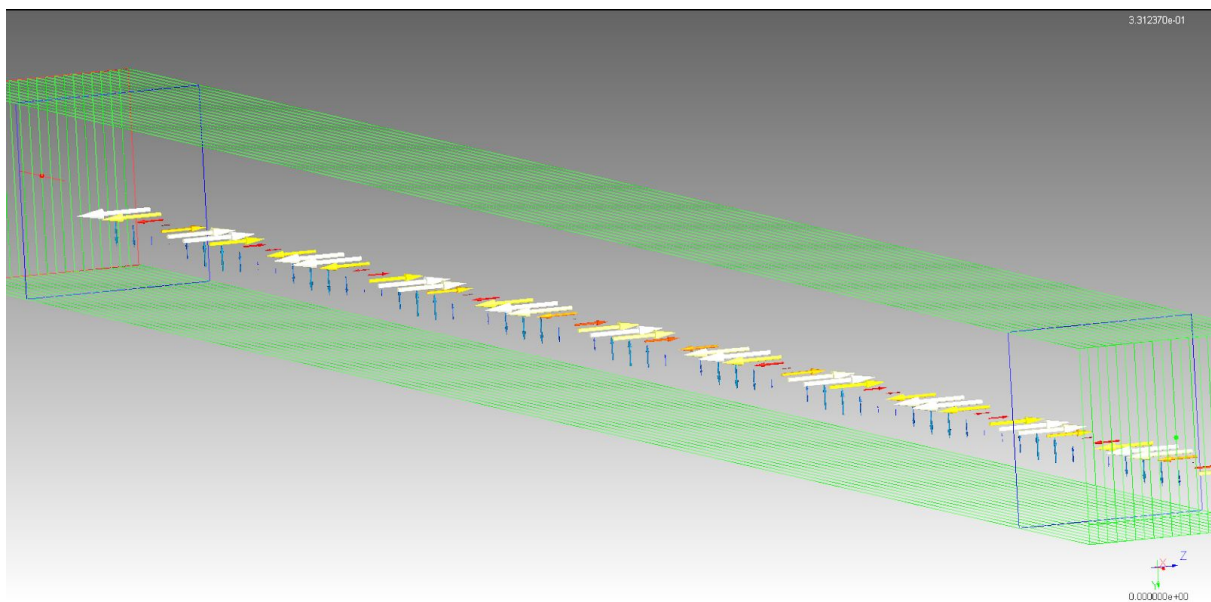


Figure 10: Fields window (3D vector display)

$f = 7.5 \text{ [GHz]}$ ,  $\epsilon_r = 1 \text{ [F / m]}$ ,  $\mu_r = 7.5 \text{ [H / m]}$ ,  $\text{tg}\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$  (*lossless model*)

Electric component: z (We can see that only z axis is visible in electric component in the program)

Magnetic component: y (We can see that only y axis is visible in magnetic component in the program )

Direction of propagation: x (It is the only axis left, after we established that we can see z and y component)

3.9

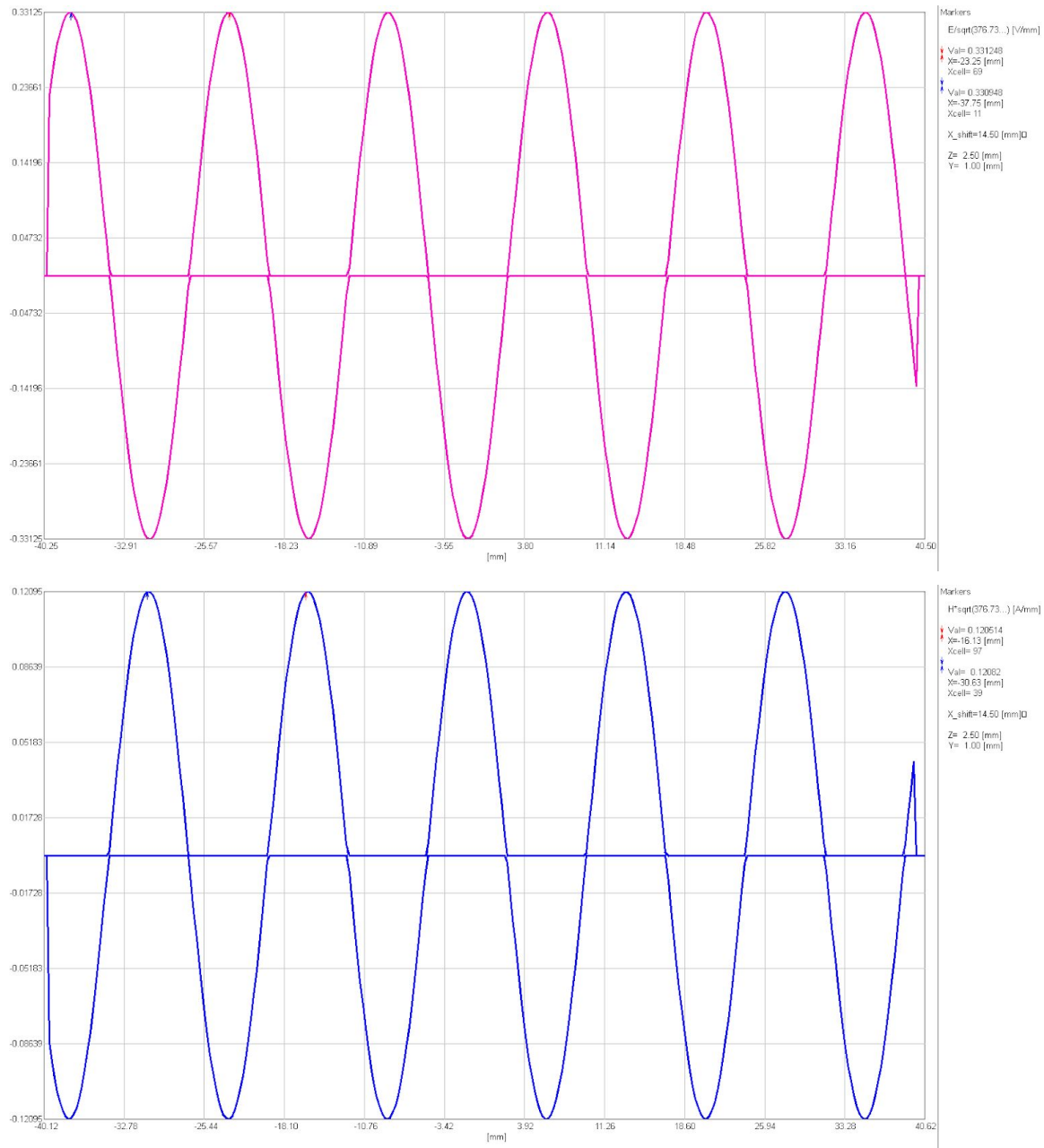


Figure 11: Envelope windows  $E_z$ (upper) and  $H_y$ (lower)

$f = 7.5 \text{ [GHz]}$ ,  $\epsilon_r = 1 \text{ [F / m]}$ ,  $\mu_r = 7.5 \text{ [H / m]}$ ,  $\text{tg}\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$  (lossless model)

Wavelength -  $\lambda = X\_shift = 14.50 \text{ [mm]}$

Formula for phase coefficient  $\beta$  using measured  $\lambda$  :

$$\lambda = \frac{2\pi}{\beta} \Rightarrow \beta = \frac{2\pi}{\lambda} \Rightarrow \beta \approx 433.3 \text{ [1 / m]}$$

$$\epsilon = \epsilon_0 * \epsilon_r$$

$$\text{Analytical formula for } \beta = \omega * \sqrt{\mu * \epsilon} = 2\pi * f * \sqrt{\mu * \epsilon} \approx 394.98 \text{ [1 / m]}$$

$\beta_{\text{markers}}$  - Phase coefficient calculated from lambda from markers

$\beta_{\text{analytical}}$  - Phase coefficient calculated from analytical formula

$$\text{Relative error} = 100 \% * \frac{\beta_{\text{markers}} - \beta_{\text{analytical}}}{\beta_{\text{analytical}}} \approx 9.7 \%$$

From markers:

$$E_n = 0.33 \text{ [V / mm]}$$

$$H_n = 0.12 [A / mm]$$

$$E = E_n * \sqrt{Z_0} \approx 6.4 [V / mm]$$

$$H = \frac{H_n}{\sqrt{Z_0}} = 0.006 [A / mm]$$

$$Z_w = \frac{E_n}{H_n} * Z_0 = 137 [\Omega]$$

From analytical formulas:

$$Z = \sqrt{\frac{\mu}{\epsilon}} = 148 [\Omega]$$

$$Relative\ error = 100\% * \frac{Z - Z_w}{Z} \approx 7.4\%$$

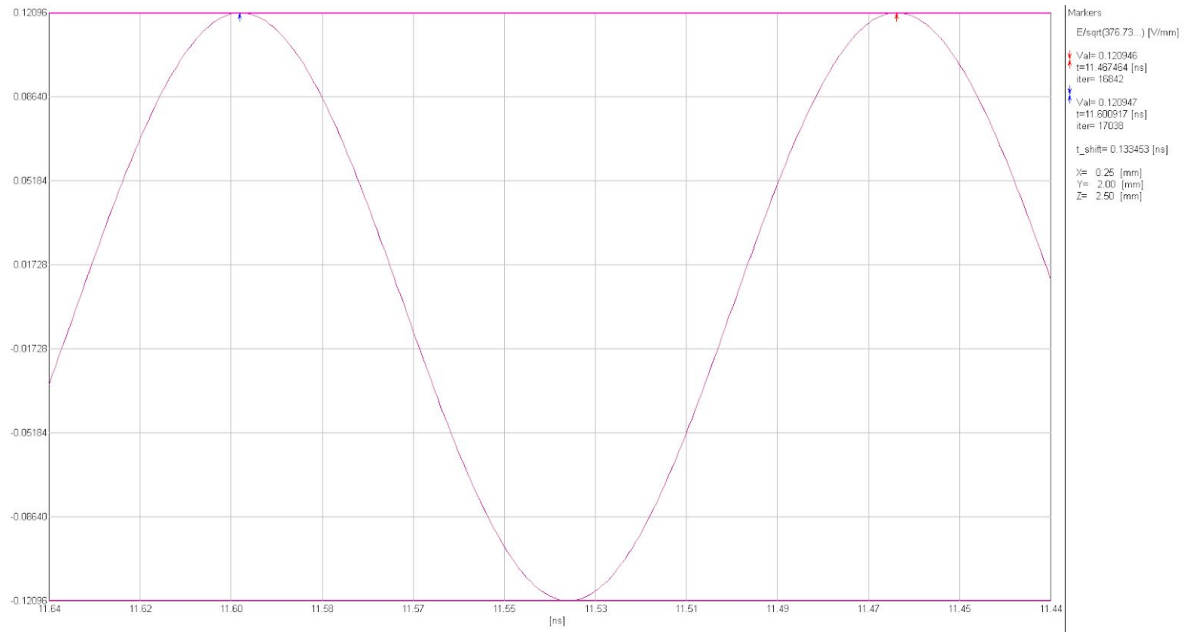


Figure 12: View envelope window for

$f = 7.5 [GHz]$ ,  $\epsilon_r = 1 [F / m]$ ,  $\mu_r = 7.5 [H / m]$ ,  $tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$  (lossless model)

$$T = t_{shift} \approx 0.1335 [ns]$$

$$T_{real} = \frac{1}{f_{real}} \approx 0.133 [ns]$$

$$Relative\ error = \frac{T - T_{real}}{T_{real}} * 100\% = 0.05\%$$

$$f = \frac{1}{T} = 7.49 [GHz]$$

$$f_{real} = 15 [GHz]$$

$$Relative\ error = \frac{f_{real} - f}{f_{real}} * 100\% = 0.13\%$$

$$\beta = 2\pi * f * \sqrt{\mu * \epsilon} \approx 394.4 [1 / m]$$

$$\beta_{analytical} = 394.98 [1 / m]$$

$$Relative\ error = \frac{\beta_{analytical} - \beta}{\beta_{analytical}} * 100\% = 0.15\%$$

$\beta$  compared with  $\beta$  from 3.7 :

$$Relative\ error\ with\ \beta\ from\ 3.7 : \frac{\beta_{3.7} - \beta_{3.10}}{\beta_{3.7}} * 100\% \approx 8.98\%$$

2 a)

Values we start with:

$f = 7.5 [GHz]$ ,  $\epsilon_r = 7.5 [F / m]$ ,  $\mu_r = 1 [H / m]$ ,  $tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$  (lossless model)

$f$  – frequency,  $\epsilon_r$  – permittivity,  $\mu_r$  – permeability,  $tg\delta$  – loss tangent,  $\sigma$  – conductivity

3.7:

$$Z_c \text{ of input} = Z_c \text{ of output} = 137.5625 [\Omega]$$

$Z_c$  – impedance

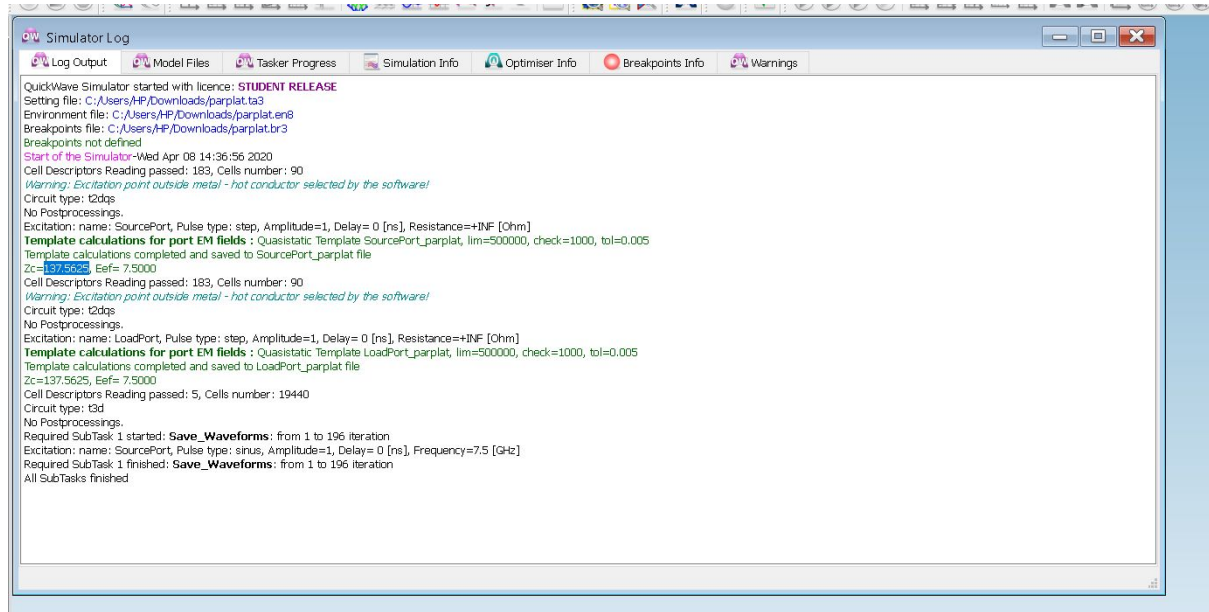


Figure 13: Simulation log window for:

$$f = 7.5 [GHz], \epsilon_r = 1 [F / m], \mu_r = 7.5 [H / m], tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$$

3.8:

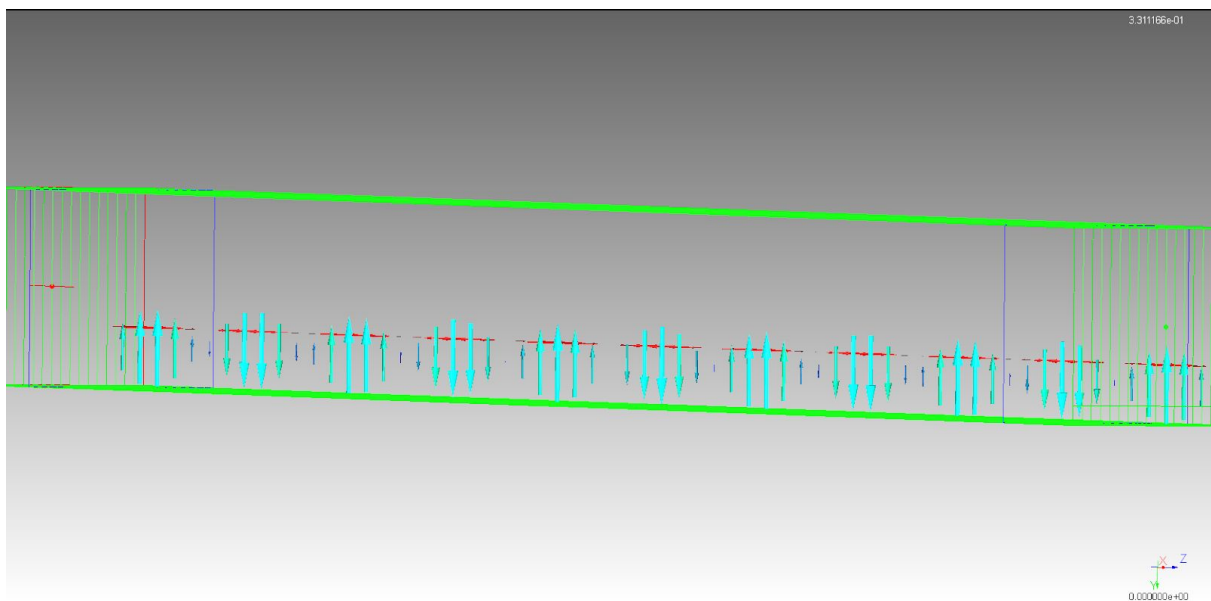


Figure 14: View Fields window (3D vector display) for:

$f = 7.5 \text{ [GHz]}$ ,  $\epsilon_r = 1 \text{ [F / m]}$ ,  $\mu_r = 7.5 \text{ [H / m]}$ ,  $\text{tg}\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$

Electric component: z (We can see that only z axis is visible in electric component in the program)

Magnetic component: y (We can see that only y axis is visible in magnetic component in the program )

Direction of propagation: x (It is the only axis left, after we established that we can see z and y component)

3.9

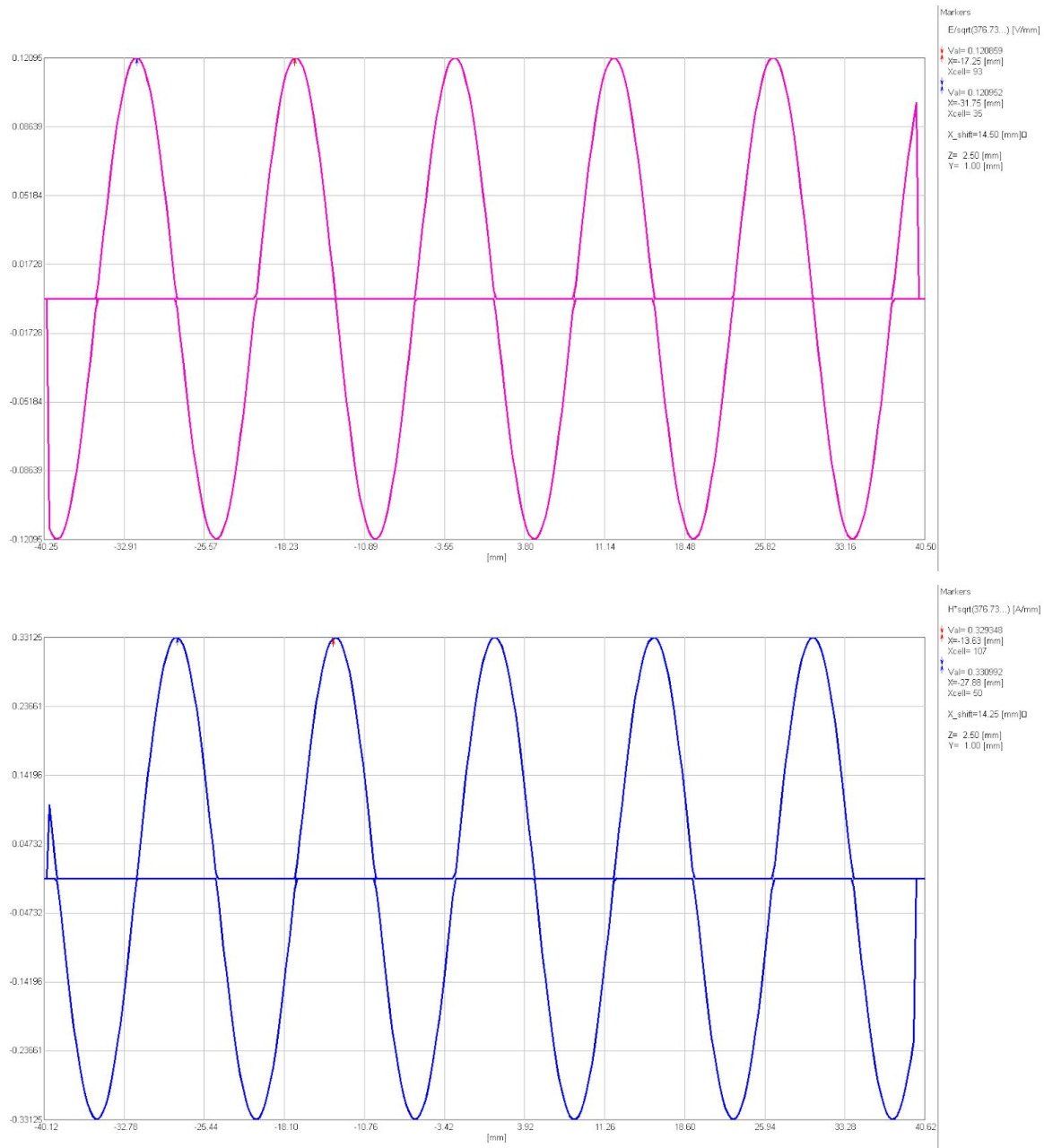


Figure 15: View envelope window for:

$$f = 7.5 \text{ [GHz]}, \epsilon_r = 1 \text{ [F / m]}, \mu_r = 7.5 \text{ [H / m]}, \text{tg}\delta = 0 \Rightarrow \sigma = 0 \text{ [S / m]}$$

$$\text{Wavelength} - \lambda = X\_shift = 14.50 \text{ [mm]}$$

Formula for phase coefficient  $\beta$  using measured  $\lambda$  :

$$\lambda = \frac{2\pi}{\beta} \Rightarrow \beta = \frac{2\pi}{\lambda} \Rightarrow \beta \approx 433.3 \text{ [1 / m]}$$

$$\epsilon = \epsilon_0 * \epsilon_r$$

$$\text{Analytical formula for } \beta = \omega * \sqrt{\mu * \epsilon} = 2\pi * f * \sqrt{\mu * \epsilon} \approx 394.98 \text{ [1 / m]}$$

$\beta_{\text{markers}}$  - Phase coefficient calculated from lambda from markers

$\beta_{\text{analytical}}$  - Phase coefficient calculated from analytical formula

$$\text{Relative error} = 100 \% * \frac{\beta_{\text{markers}} - \beta_{\text{analytical}}}{\beta_{\text{analytical}}} \approx 9.7 \%$$

From markers:



$$E_n = 0.12 [V / mm]$$

$$H_n = 0.33 [A / mm]$$

$$E = E_n * \sqrt{Z_0} \approx 2.33 [V / mm]$$

$$H = \frac{H_n}{\sqrt{Z_0}} = 0.017 [A / mm]$$

$$Z_w = \frac{E_n}{H_n} * Z_0 = 137 [\Omega]$$

From analytical formulas:

$$Z = \sqrt{\frac{\mu}{\epsilon}} = 148 [\Omega]$$

$$\text{Relative error} = 100\% * \frac{Z - Z_w}{Z} \approx 7.4 \%$$

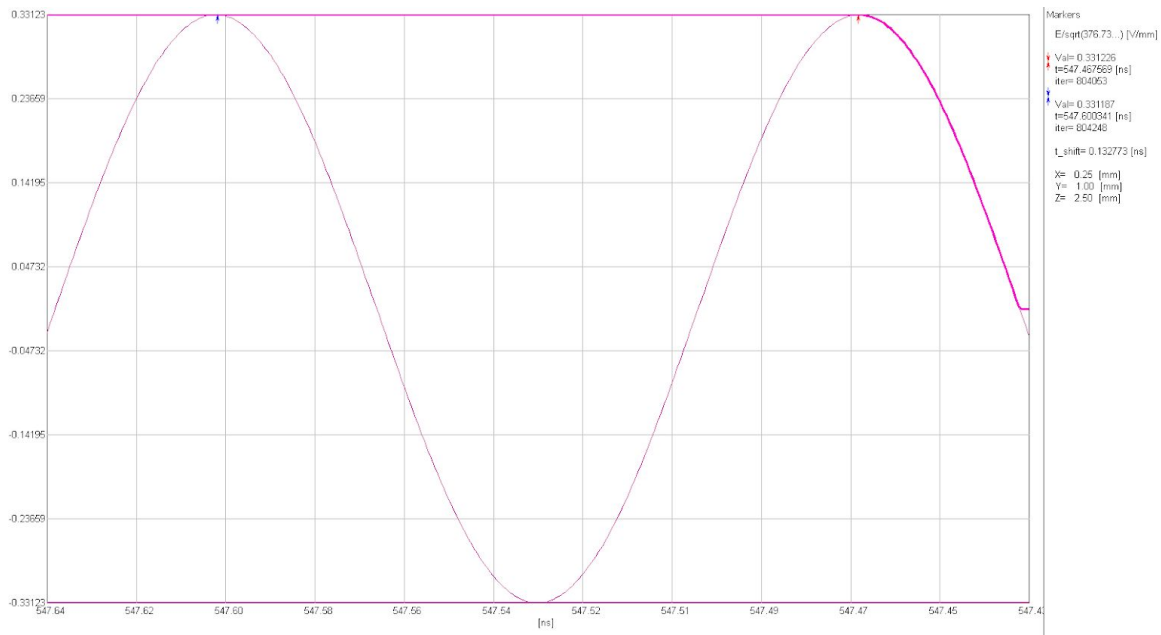


Figure 16: Time domain View Envelope window for:

$$f = 7.5 [GHz], \epsilon_r = 1 [F / m], \mu_r = 7.5 [H / m], tg\delta = 0 \Rightarrow \sigma = 0 [S / m]$$

$$T = t_{shift} \approx 0.1327 [ns]$$

$$T_{real} = \frac{1}{f_{real}} \approx 0.133 [ns]$$

$$\text{Relative error} = \frac{T - T_{real}}{T_{real}} * 100\% = 0.23\%$$

$$f = \frac{1}{T} = 7.53 [GHz]$$

$$f_{real} = 7.5 [GHz]$$

$$\text{Relative error} = \frac{f_{real} - f}{f_{real}} * 100\% = 0.4\%$$

$$\beta = 2\pi * f * \sqrt{\mu * \epsilon} \approx 396.6 [1 / m]$$

$$\beta_{analytical} = 394.98 [1 / m]$$

$$\text{Relative error} = \frac{\beta_{analytical} - \beta}{\beta_{analytical}} * 100\% = 0.41\%$$

$\beta$  compared with  $\beta$  from 3.7 :

$$\text{Relative error with } \beta \text{ from 3.7 : } \frac{\beta_{3.7} - \beta_{3.10}}{\beta_{3.7}} * 100\% \approx 0.41\%$$

Answering the questions at the end:

a) frequency is proportional with  $x_{shift}$

- b) permittivity and permeability changes E and H, but frequency and  $x_{\text{shift}}$  does not change
- c) it is a sinus shape
- d) It is  $\frac{\lambda}{2}$